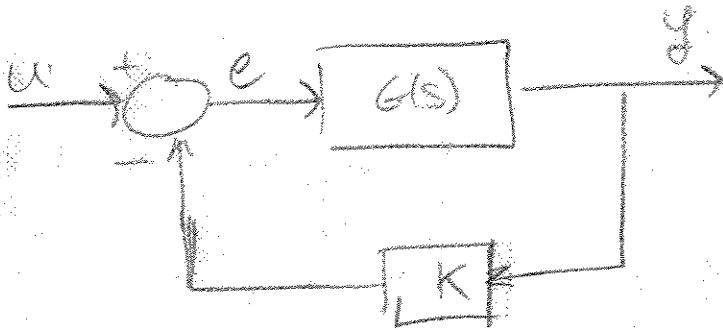


# Root Locus

Consider the feedback system.



where  $G(s) = \frac{n(s)}{d(s)}$  and  $\deg(n(s)) < \deg(d(s))$ .

The closed-loop transfer function for this system is given by,

$$H(s) = \frac{G(s)}{1 + KG(s)} = \frac{n(s)/d(s)}{1 + K n(s)/d(s)}$$

$$= \frac{n(s)}{d(s) + Kn(s)}$$

We are interested in how  $K$  affects the CL poles as  $K$  varies from 0 to  $\infty$ .

i.e. the CL-poles are the roots of the equation

$$\boxed{d(s) + Kn(s) = 0}$$

Def. The locus of these roots plotted in the complex plane  $\mathbb{C}$  as a function of  $K$  is called the root-locus, and  $0 < K < \infty$ .

$$d(s) + Kn(s) = 0$$

If  $K=0$ , the equation becomes  $d(s) = 0$ .

$\therefore$  The locus starts at the poles of  $G(s)$ .

Note that we have  $\frac{1}{K} d(s) + n(s) = 0$ .

Then as  $K \rightarrow \infty$ , the equation becomes  $n(s) = 0$ .

$\therefore$  The locus ends at the zeros of  $G(s)$ .

# Rules for Graphical Construction

1) OL Poles and Zeros (assume  $n$  poles,  $M$  zeros,  $n > M$ )  
Mark the  $n$  OL poles with an "x"  
Mark the  $M$  OL zeros with an "o"

2) Branches of the Root Locus:

The root locus has exactly  $n$  branches, each branch of these,  $M$  branches will approach the  $M$  OL zeros as  $k \rightarrow \infty$ .

The remaining  $n - M$  branches will go to infinity.  
(Note:  $n$  branches b/c it is the locus of a polynomial of degree  $\deg(d(s)) = n$  as a parameter,  $k$ , varies)

3) Asymptotic Rays:

The  $n - M$  branches going to infinity asymptotically approach  $n - M$  rays originating at a single point on the real axis, these rays are the asymptotes and their common origin is the center of the asymptotes.

The angle of the asymptotes are given by

$$\alpha = \frac{2l+1}{n-m} \pi, \quad l=0, 1, 2, \dots, n-m-1.$$

i.e.

$n-m$	$\alpha$
1	$\pi$
2	$\pm \pi/2$
3	$\pm \pi/3, \pi$
4	$\pm \pi/4, \pm 3\pi/4$

The center of the asymptotes is given by

$$\sigma_c = \frac{\sum p_j - \sum z_i}{n-m}$$

$p_j$  = poles

$z_i$  = zeros

of OL system.

#### 4) Real Part of the Root Locus

The part of the real axis to the left of an odd number of poles or zeros belong to the root locus.

### 5) Breakaway Points:

These are points at which two (or more) branches of the root-locus depart or arrive at the real axis.



How to compute a breakaway point?

Recall,  $d(s) + kn(s) = 0$ .

Then  $K(s) = -\frac{d(s)}{n(s)}$ .

a breakaway point,  $\sigma_b$ ,  $K(s)$  obtains a maximum.

Thus, to find  $\sigma_b$ , we need to solve,

$$\underline{\underline{\frac{dK(s)}{ds} = 0}}$$

Alternative approach: solve the equation,

$$\sum_{i=1}^n \frac{1}{(\sigma_b + p_i)} = \sum_{i=1}^m \frac{1}{(\sigma_b + z_i)}$$

$\sigma_b$  is the breakaway point.

### c) Crossover Points:

These are the points where the branches of the root locus cross the imaginary axis

→ found by Routh-Hurwitz

Example: plot the root locus for

$$G(s) = \frac{1}{(s+1)(s+2)(s+3)} = \frac{n(s)}{d(s)}$$

The CL pole equation is  $d(s) + K n(s) = 0$ :

$$s^3 + 6s^2 + 11s + (6 + K) = 0$$

OL poles are:  $p_1 = -1, p_2 = -2, p_3 = -3$

No OL zeros

$$\Rightarrow n=3, m=0 \Rightarrow n-m=3$$

∴ the RL has 3 branches, all going to infinity along the asymptotes centred at the breakaway point.

$$\sigma_c = \frac{\sum p_i - \sum z_i}{n-m} = \frac{-1-2-3}{3} = \boxed{-2}$$

The angles of the asymptotes are given by

$$\alpha = \frac{2l+1}{3} \pi, \quad l = 0, 1, 2$$

$l=0 \Rightarrow \alpha = \frac{\pi}{3} = 60^\circ$
$l=1 \Rightarrow \alpha = \pi = 180^\circ$
$l=2 \Rightarrow \alpha = \frac{5\pi}{3} = 300^\circ$

The intervals  $(-\infty, -3]$  and  $[-2, -1]$  are the real part of the RL since they are to the left of an odd number of poles.

The breakaway point is computed using

$$K(s) = \frac{-d(s)}{n(s)} = -(s^3 + 6s^2 + 11s + 6)$$

$$\Rightarrow \frac{dK(s)}{ds} = -3s^2 - 12s - 11 = 0.$$

$$\Rightarrow s = -2 \pm \frac{\sqrt{3}}{3}$$

Choose:  $s = -2 + \frac{\sqrt{3}}{3}$  since  $-2 + \frac{\sqrt{3}}{3} \in (-2, -1)$

and  $-2 - \frac{\sqrt{3}}{3} \notin (-2, -1)$ .

Use Routh-Hurwitz table to find crossover points:

$$s^3 + 6s^2 + 11s + (6+K) = 0.$$

$$\begin{array}{c|cc} s^3 & 1 & 11 \\ s^2 & 6 & 6+K \\ s^1 & b_1 = 10 - \frac{K}{6} & 0 \\ s^0 & 6+K & 0 \end{array}$$

$$\begin{aligned} b_1 &= -\frac{1}{6} \begin{vmatrix} 1 & 11 \\ 6 & 6+K \end{vmatrix} \\ &= -\frac{1}{6} (6+K - 66) \\ &= 10 - \frac{K}{6} \end{aligned}$$

We need  $10 - \frac{K}{6} > 0$  and  $6+K > 0$

$$\Rightarrow K < 60 \quad \text{and} \quad K > -6$$

only need  $K < 60$  since  $0 < K < 10$ .

So our crossover point is  $K_c = 60$

